Research



Assessment of Effects of a Delay Block and a Nonlinear Block in Systems with Chaotic Behavior Using Lyapunov Exponents

Evaluación del Efecto de un Retardo y una No Linealidad en Sistemas con Comportamiento Caótico Utilizando Exponentes de Lyapunov

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Abstract

Context: Because feedback systems are very common and widely used, studies of the structural characteristics under which chaotic behavior is generated have been developed. These can be separated into a nonlinear system and a linear system at least of the third order. Methods such as the descriptive function have been used for analysis.

Method: A feedback system is proposed comprising a linear system, a nonlinear system and a delay block, in order to assess his behavior using Lyapunov exponents. It is evaluated with three different linear systems, different delay values and different values for parameters of nonlinear characteristic, aiming to reach chaotic behavior.

Results: One hundred experiments were carried out for each of the three linear systems, by changing the value of some parameters, assessing their influence on the dynamics of the system. Contour plots that relate these parameters to the Largest Lyapunov exponent were obtained and analyzed.

Conclusions: In spite non-linearity is a condition for the existence of chaos, this does not imply that any nonlinear characteristic generates a chaotic system, it is reflected by the contour plots showing the transitions between chaotic and no chaotic behavior of the feedback system.

Keywords: Chaos, Dynamical Systems, Lyapunov Exponents, Nonlinear Characteristic.

Language: English.



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Resumen

Contexto: Al ser los sistemas realimentados muy comunes y ampliamente usados, se han desarrollado estudios de las características estructurales bajo las cuales se genera comportamiento caótico. Estos pueden ser separados en un sistema no lineal y un sistema lineal por lo menos de tercer orden. Se han usado métodos como la función descriptiva para su análisis.

Método: Se propone un sistema realimentado a partir de un sistema lineal, un sistema no lineal y un retardo, con el fin de evaluar su comportamiento utilizando los exponentes de Lyapunov. Se evalúa con tres diferentes sistemas lineales, diferentes valores del retardo y diferentes valores para los parámetros de una característica no lineal, buscando alcanzar un comportamiento caótico.

Resultados: Se realizaron cien experimentos para cada uno de los tres sistemas cambiando el valor de algunos parámetros evaluando la influencia de los mismos en la dinámica del sistema. Se realizan y analizan gráficas de contorno que relacionan estos parámetros con el máximo exponente de Lyapunov.

Conclusiones: A pesar que la no linealidad es una condición para que exista caos, esto no implica que cualquier característica no lineal genera un sistema caótico, esto se evidencia en las gráficas de contorno mostrando las transiciones entre comportamiento caótico y no caótico del sistema realimentado.

Palabras clave: Caos, Característica No lineal, Exponentes Lyapunov, Sistemas Dinámicos.

Idioma: Inglés

1. Introduction

The chaos has been object of study since past century and it has generated more interest in the researchers since was discovered that in nature exist systems with this type of behavior [1]. The chaotic systems also have been object of study in the control area due to they can be analyzed with mathematical tools of dynamical systems [2]. In practice is common to see systems which exhibit pure delays, either by their nature or by the controller used, anyway the system behavior is affected. One particular case are the systems with negative feedback which exhibit chaotic behavior under specific structural conditions. In [3], [4] feedback systems are shown which exhibit this behavior, also a structure is shown which divides the systems into a linear block and a nonlinear block in order to facilitate the analysis. It enlarges the study scope to search structural conditions and pure delay effects which generate chaotic behavior, which can be evaluated via different methods. For example, in [4]–[6] an evaluation is made by the descriptive function method. In [7], dynamical systems with time-delayed feedback are studied using Lyapunov exponents to identify strong or weak chaos. In [8], a numerical analysis of chaotic behavior of fractional-order Chen system is performed, using Lyapunov exponents and bifurcations diagrams.

In this work the chaotic behavior of a feedback system proposed is evaluated using Lyapunov exponents and phase plane. This system is composed of a linear block, a nonlinear block and a delay. Three linear systems are proposed, a pure delay which is implemented in simulation using the *Transport Delay* Simulink block and the non-linearity is built. In contrast to works [7], [8], the initial point is not a known chaotic system. In this case, the chaotic system is designed, built and assessed in order to identify which values of parameters are most influentials to chaotic behavior.

This paper is organized as follows: In sections 2, 3 and 4 is described the conceptual framework for develop the study, in section 5 the proposed feedback system and its elements are described

in greater detail, in section 6 is showed how the simulation was implemented and the outcomes obtained for each evaluated case, and in section 7 the conclusions obtained are described.

2. Delay Systems

A continuous system is the one in which the continuous input signals are transformed into continuous output signals. A system can present an inherent delay that causes a time-shift in the input signal but does not affect his characteristics, this kind of delay is called pure delay. In practice is possible to find pure delays in some kind of systems, especially those with hydraulic, pneumatic or mechanical transmissions. The computer control systems also have delays as they require some time to perform numerical operations [9].

2.1. Padé Approximation

The pure delay transfer function is $e^{-L \cdot S}$, which Padé approximation is a rational function (1), where $P_m(s)$ is a polynomial defined by (2) and $Q_n(s)$ is a polynomial defined by (3) where n and m represent the degree of each polynomial [10]. The roots of polynomial $P_m(s)$ are called zeros and the roots of polynomial $Q_n(s)$ are called poles [9].

$$e^{-L \cdot S} = \frac{P_m(S)}{Q_n(S)} \tag{1}$$

$$P_m(S) = \sum_{k=0}^{m} \frac{(m+n-k)!m!}{(m+n)!k!(m-n)!} (-S)^k$$
 (2)

$$Q_n(S) = \sum_{k=0}^n \frac{(m+n-k)!n!}{(m+n)!k!(n-k)!} (-S)^k$$
(3)

Is possible to choose n=m to add the same number of poles and zeros. This approximation is used to know what order generates a similar behavior to feedback system with delay.

3. Simple Function

A simple function is a function that satisfies $\varphi: A \subseteq \mathbb{R} \to \mathbb{R}$ which is given by:

$$\varphi(x) = \sum_{i=1}^{n} g(x)\chi_{A_i}(x) \tag{4}$$

where $\chi_{A_i}(x)$ is the i-th indicator function [11].

if $g(x) = c \ \forall \ x \in A$, φ is called zero-order.

if $g(x) = c_0 + c_1 x \ \forall \ x \in A, \varphi$ is called first-order.

where $c, c_0, c_1 \in \mathbb{R}$ and are constants.

Below the simple function of first-order is used. This function is given by (5), which is used because of the ease to change the simulation parameters.

$$\varphi(x) = \sum_{i=1}^{n} (c_0 + c_1 x) \chi_{A_i}(x)$$
(5)

4. An Approach to Chaos in Dynamical Systems

A universally accepted definition for chaos does not exist yet, but almost every authors agree in three main features [12]:

- 1. *Aperiodic long-term behavior:* There are trajectories which do not settle down to fixed points, periodic orbits, or quasiperiodic orbits in the course of time.
- 2. *Deterministic:* The irregular behavior arises from the non-linearity of the system, rather than from noisy driving forces.
- 3. Sensitive dependence on initial conditions: Two different initial conditions but nearby each other, generate different trajectories.

Chaos can appear in feedback systems and a way to assess his behavior is through of Lyapunov exponents.

4.1. Feedback Systems

Due to the frequent use of feedback systems and all tools that exist for its analysis, it has been studied under which structural conditions a chaotic behavior is presented.

In [4] is suggested that a feedback system presents a chaotic behavior when a predicted limit cycle (PLC) and an equilibrium point (EP) of certain characteristics interact between themselves.

For chaos to be presented must exist a non-linearity in the system, therefore the feedback system can be represented as a linear subsystem and a nonlinear subsystem, also the lineal subsystem must be at least third order (Figure 1).

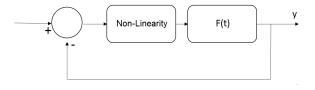


Figure 1. Structure feedback system (Source: [4])

4.2. Lyapunov Exponents

Two trajectories are considered which are separated by an initial distance d_0 at time t_0 that present a growth of distance (d_t) between them, which is calculated by (6) over a time period Δt [13], as illustrated in Figure 2.

$$d_t = d_0 e^{\lambda \Delta t} \tag{6}$$

$$\lambda = \frac{1}{\Delta t} \ln(\frac{d_t}{d_0}) \tag{7}$$

The number λ is called Lyapunov exponent, is defined by (7) and is used to quantify the chaotic behavior in dynamical systems. The authors in [14], [15], use Lyapunov exponents to study the behavior of a prototype electrochemical oscillator and the magnetic anisotropy effects in a magnetic particle respectively.

Depending on the value of the Lyapunov exponent there are three options:

- $\lambda < 0$, implies contraction, it means that the system converges to a fixed point.
- $\lambda = 0$, the system tends to a limit cycle.
- $\lambda > 0$, implies distance growth between the trajectories, the system presents chaotic behavior.

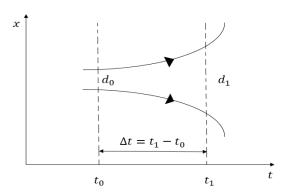


Figure 2. Two nearby divergent trajectories (Source: [13])

There are different methods to calculate the Lyapunov exponents as shown in [16], [17]. In the realized tests of this work the Gram-Schmidt method is used with a time span of 100 units time and 0.01 units time of transient time to three second order systems with delay. This method uses orthonormal initial conditions, e.g. $u_1 = [1, 0, 0, ...], u_2 = [0, 1, 0, ...], ...$ Variables w_1, v_1 y w_2 are calculated; Then, using w_2 value and equation 8 the Largest Lyapunov exponent is obtained, where K is the number of samples and T is the sample period.

$$w_{1} = u_{1}, v_{1} = w_{1} / \|w_{1}\|$$

$$w_{2} = u_{2} - \langle u_{2}, v_{1} \rangle v_{1}, v_{2} = w_{2} / \|w_{2}\|$$

$$\lambda_{2} = \frac{1}{KT} \sum_{i=1}^{K} \ln \|w_{2}^{i}\|$$
(8)

5. System Description

In practice, pure delays are presented in several kinds of systems [9], for this reason, a study is performed to observe whether chaotic behavior is generated from a linear system with delay and a non-linearity. One feedback system was built with these elements, as can see in Figure (3).

To perform the study, a linear system and a non-linearity were proposed following the structure suggested in [4] (Figure 1), the difference is in linear system by second order with delay and the non-linearity is built. Below explain with detail each part of the proposal.

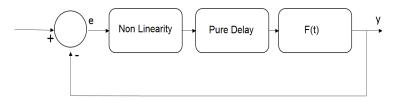


Figure 3. Feedback system structure proposed (Source: Authors)

5.1. Linear System

As is proposed in [4], a dynamical system is chaotic if his linear part is a oscillatory system and exist some feedback control features which make that system to present at least a predictable limit cycle. According to the above, three linear systems are proposed to analyze several cases or scenarios.

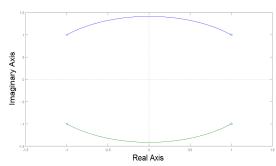
First System: It has a couple of complex conjugated poles with negative real part and two complex conjugate zeros with positive real part, which are symmetrical respect to imaginary axis. In Figure 4 is shown the root locus with the characteristics described.

The system transfer function is

$$G(S) = \frac{s^2 - 2s + 1}{s^2 + 2s + 1} \tag{9}$$

The state variables representation is

$$\dot{x_1} = x_2
\dot{x_2} = -2x_1 - 2x_2 + u
y = 2x_1 - 2x_2 + \dot{x_2}$$



(10) **Figure 4.** Root locus of the first linear system (Source: Authors)

Second System: It has a couple of complex conjugated zeros with positive real part and two real negative repeated poles. In Figure 5 is shown the root locus with the characteristics described.

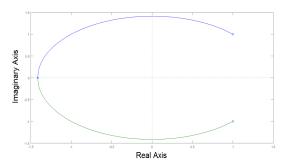
The system transfer function is

$$G(S) = \frac{s^2 - 2s + 2}{s^2 + 2,828s + 2}$$
 (11)

The state variables representation is

$$\dot{x_1} = x_2$$

 $\dot{x_2} = -2x_1 - 2,828x_2 + u$
 $y = -4,828x_2 + u$



(12) **Figure 5.** Root locus of the second linear system (Source: Authors)

Third System: It has a couple of complex conjugated zeros with positive real part and two real negative different poles. In Figure 6 is shown the root locus with the characteristics described.

The system transfer function is

$$G(S) = \frac{s^2 - 2s + 2}{s^2 + 2,828s + 1,75}$$
 (13)

The state variables representation is

$$\dot{x_1} = x_2
\dot{x_2} = -1,75x_1 - 2,828x_2 + u
y = 0,25x_1 - 4,828x_2 + u$$
(14)

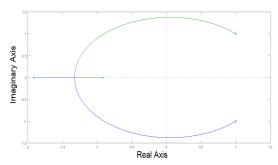


Figure 6. Root locus of the third linear system (Source: Authors)

5.2. Nonlinear Block

For this particular case, the nonlinear characteristic is not inherent to the system, so it is proposed a nonlinear block built from linear functions piecewise, and is implemented using the first order simple function defined by (5).

The nonlinear characteristic built is defined by (15) and is shown in Figure 7. The variable e is the input and represents the system error, defined by e = -y being "y" the feedback system output; the variable u is the output of the nonlinear characteristic and represents the input to the linear system with delay.

$$u(e) = \begin{cases} m_1 e - b & , if & e < -h \\ m_2 e & , if & -h < e < h \\ m_1 e + b & , if & e > h \end{cases}$$
 (15)

In Figure 7 the parameters that constitute the nonlinear characteristic are indicated. These Parameters are:

- m_1 : Slope of linear equations of the ends
- m_2 : Slope of internal linear equation
- b: Vertical axis intercept of the linear equations with slope m_1
- h: It represents the transition between a linear equation and other.

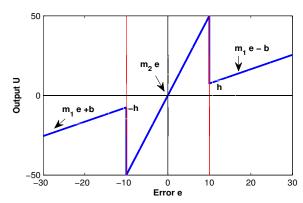


Figure 7. Proposed nonlinear characteristic (Source: Authors)

Some considerations taken in order to select each parameter are the following. First of all, based on root locus of each linear system, the gain value where the systems are marginally stable was found and is called k_m , for each system the value was $k_m = 1$. Based on this gain, the parameters

 m_1 y m_2 were selected.

The gain value m_2 must be greater than k_m to move the poles to the unstable region and the gain value m_1 must be less than k_m to bring back the poles toward the stable region. When the error is into the m_2 zone, the system becomes unstable and the error value increases until it reaches the m_1 zone where the system becomes stable. These gain changes make the system have a erratic behavior and therefore, uncertainty of new values that can take the error is generated and it opens the possibility of a chaotic behavior in the system. Values of m_1 and m_2 that generate stronger changes of the poles position may exist. By last, the parameters h and h are selected arbitrarily. In the results section a tuning of all parameters described previously is made for a particular system, in order to identify what combinations generate chaotic behavior.

5.3. Delay

When a delay is analyzed with the Padé approximant described in section 2.1, the delay adds infinite poles and zeros to the linear system and it is converted in a superior order system. With this, the condition proposed in [4] is achieved, where is suggested that the linear system should be third order or greater. Different delay values were tried (since 0.5 until 7 time units) for a particular linear system in order to analyze effects in the feedback system.

6. Results

This section shows the implementation of the feedback system simulation which was proposed and described in preceding section. The parameters m_1 and b are fixed and the parameters m_2 , h and the delay time (t_d) are modified in order to evaluate their effects in the behavior of three feedback systems.

6.1. Simulation

The state equations of the three linear systems are implemented (10, 12, 14). The *Transport Delay* Simulink block is used for set the delay, and finally for the nonlinear characteristic three indicator functions are created, one for each piece of the equation (15) and using the first order simple function defined by (5) the straight equations are implemented, this implementation is shown in Figure (8).

To assess the chaotic behavior of the feedback system, a copy is created with initial conditions different but very close to the original conditions and orthonormal between them $(u_1 = \{x_1 = 1, x_2 = 0\}, u_2 = \{x_1 = 0, x_2 = 1\})$. The data of each state variable for a time span of 100 time units are collected and exported, and the Largest Lyapunov exponents are calculated using (8) with K = 10000 and a sample period T = 0.01 time units.

6.2. Tests with Different Linear Systems

Following the considerations described in section 5, a previous study of all parameters $(m_1, m_2, h, b \ y \ t_d)$ was made, obtaining that the parameters m_2 , $h \ y \ t_d$ are the most influential in

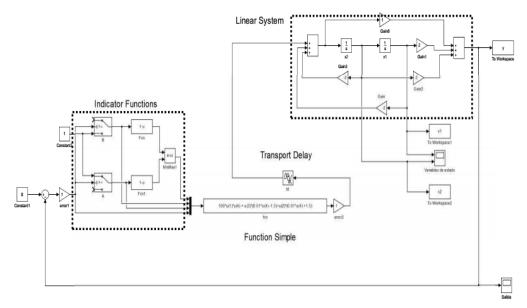


Figure 8. Simulation of the second order system with delay and nonlinear characteristic in negative feedback.

the feedback system behavior. In the feedback system, the linear system is changed by the three systems proposed and 100 tests are made to each one, changing the parameters values in this way: m_2 between 1.1-101.1, h between 1-4 and t_d between 0.5-7. For each parameter combination, the Largest Lyapunov exponent is computed with the equation 8 and the contour plots that show how changes this exponent in function of the three parameters are showed in figures 9, 11 and 13. When the Largest Lyapunov exponent is least or equal to zero it is codified in black color and it means that there is not chaotic behavior. Also, in those figures the transitions between chaotic and no chaotic behavior are shown.

First System:

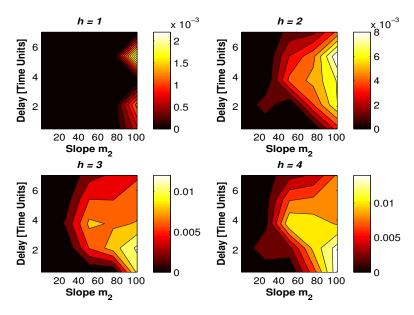


Figure 9. Contour plots of the Largest Lyapunov exponent color-coded as a function of m_2 , t_d and h to the first linear system. The fixed parameters are $m_1=0.01$ and b=1.5. The resolutions are : $\Delta m_2=25$, $\Delta t_d=1.625$ and $\Delta h=1$.

As shown in figure 9, as h value increases, the zone where the Largest Lyapunov exponent is positive concentrates in the lower right corner, it means t_d values between 0.5 and 2.125, and $m_2 = 101.1$. The equation 8 is used to calculate the Largest Lyapunov Exponent of the 100 tests with the first linear system, and the highest was 0.01553 with h = 4, $m_2 = 101.1$ and $t_d = 0.5$. The phase plane obtained with this values is shown in figure 10.

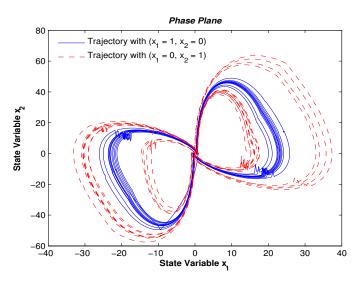


Figure 10. Phase Plane with first linear system, initial conditions $u_1 = \{x_1 = 1, x_2 = 0\}, u_2 = \{x_1 = 0, x_2 = 1\}.$

Second System:

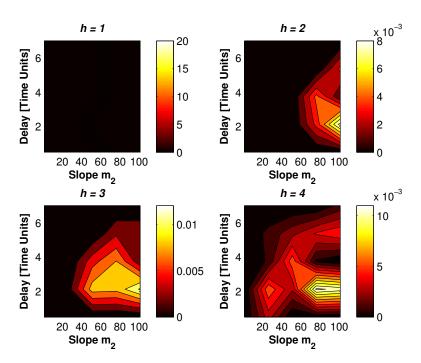


Figure 11. Contour plots of the Largest Lyapunov exponent color-coded as a function of m_2 , t_d and h to the second linear system. The fixed parameters are $m_1=0.01$ and b=1.5. The resolutions are : $\Delta m_2=25$, $\Delta t_d=1.625$ and $\Delta h=1$.

For the second linear system is highlighted that with h=1, the system behavior is not chaotic to any combination of m_2 and t_d , as shown in figure 11. The positive Largest Lyapunov exponent concentrates in the zone nearby to t_d values between 2.125 and 3.75, and m_2 between 51.1 and 101.1. The equation 8 is used to calculate the Largest Lyapunov Exponent of the 100 tests with the second linear system, and the highest was 0.01124 which was obtained with h=4, $m_2=76.1$ and $t_d=2.125$. The phase plane obtained with this values is shown in figure 12.

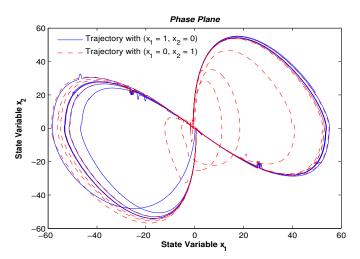


Figure 12. Phase Plane with second linear system, initial conditions $u_1 = \{x_1 = 1, x_2 = 0\}, u_2 = \{x_1 = 0, x_2 = 1\}.$

Third System:

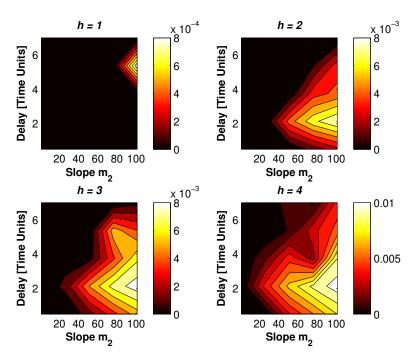


Figure 13. Contour plots of the Largest Lyapunov exponent color-coded as a function of m_2 , t_d and h to the third linear system. The fixed parameters are $m_1=0.01$ and b=1.5. The resolutions are : $\Delta m_2=25$, $\Delta t_d=1.625$ and $\Delta h=1$.

As shown in figure 13, with h=1 the Largest Lyapunov exponent concentrates in the zone nearby to $m_2=101.1$ and $t_d=5.375$. As value of h increases, the positive Largest Lyapunov exponent concentrates in the zone of $m_2=101.1$ and t_d values between 2.125 and 3.75. The equation 8 is used to calculate the Largest Lyapunov Exponent of the 100 tests with the third linear system, and the highest was 0.01056 which was obtained with h=4, $m_2=101.1$ and $t_d=2.125$. The phase plane obtained with this values is shown in figure 14.

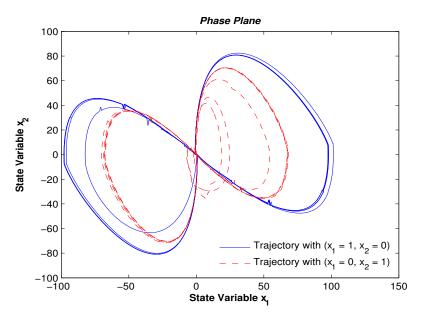


Figure 14. Phase Plane with third linear system, initial conditions $u_1 = \{x_1 = 1, x_2 = 0\}, u_2 = \{x_1 = 0, x_2 = 1\}.$

In general, changing the parameters values h, t_d and m_2 is possible to identify the transition zones between chaotic and no chaotic behavior. As h increases, the transition point between stable and unstable regions moves away from origin (Figure 7), allowing erratic behavior be more influential in the dynamical feedback system. It is manifested numerically in the increment of Largest Lyapunov exponent. For specific h and t_d values, the increment of m_2 manifests an increment of Largest Lyapunov exponent which represents the chaotic behavior in the system. The influence of delay value t_d on the Largest Lyapunov exponent in each system was particular; for the first system, the parameters t_d and h were the more influential while that in second and third system the more influential parameters were h and m_2 .

7. Conclusions

Under the considerations and the systems proposed in this paper there are certain observations.

The proposed second order systems, are more likely to have disturbances in their natural behavior due to external factors. This study allowed an assessment of the impact of a pure delay and a nonlinear block for a particular linear system, and how a built feedback system using this elements presents chaotic behavior. The tool that allowed this study was Lyapunov exponents due to they are interpretable and easy to implement when is identifying chaotic behavior in the systems.

In spite non-linearity is a condition for the existence of chaos, this does not imply that any nonlinear characteristic generates a chaotic system. Separately, delay and nonlinear characteristic make to feedback system modify its dynamic and a limit cycle is created. Together, they can generate chaotic behavior under the right conditions, such as those achieved when the parameters values were changed, watching the transitions of system dynamics between chaotic and no chaotic behavior.

This work gives rise to try out other systems with different structures, other types of non-linearity, other methods to build the non-linearity and an exploration into delay approximation with a greater order, being mindful that the chaotic behavior in the systems is being evaluated.

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